Beam-Squinting Effect in Broadband Mobile Communication

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Due to the fact that the quality of the received signal is adversely affected by the beam squint phenomenon, which is especially pertinent to the millimeter wave (mmwave) bands, many studies have been utilised by other researchers to provide light on some of the challenges that come with analysing this type of occurrence. Squint causes several issues, the most important of which are its detrimental effects on gain, line of sight, angle of arrival, progressive phase, usable bandwidth, and fading effect. As a result of these obstacles, the advantages of adopting a high-frequency band such as mmwave in modern wireless communication systems are severely limited.

beam squinting massive MIMO millimeter waves

1. Introduction

In today's world, almost everyone has a mobile device, and the number of people who do is growing. Concurrently, the number of interactive multimedia apps is also growing, and experts predict that this pattern will continue into the foreseeable future [1]. In spite of the fact that mobile technologies of the first, second, third, and fourth generations have spent the last 25 years working to increase the speed and efficiency of wireless networks, there are still some specific application areas in which current wireless networks are unable to deliver ^[2]. As the need for mobile traffic steadily develops, a bottleneck between spectrum constraints and capacity requirements is becoming more obvious. This bottleneck is becoming more and more apparent. The wireless bandwidth bottleneck is one of the most serious challenges faced by 5G communications 3. Wireless carriers need to be ready to sustain a growth in total mobile traffic that may be as much as 1000 times higher than it is now since the need for capacity in mobile broadband communications is growing at a substantial rate each year. The pursuit of inventive solutions is required in order to make the transition to the fifth generation (5G) of wireless technology ^[4]. Because of the growing need for bandwidth, the millimetre wave spectrum has been standardised for use in high-speed wireless communication ^[5]. mmwaves are an interesting possibility since they have a significant amount of bandwidth and spectrum that is readily available (approximately 60 GHz) ^[6]. mmwave communications may make use of the large capacity that is now being underutilised, which will help them fulfil the high demands of the future generation of wireless networks $[\underline{7}]$.

Although the potential bandwidth of mmwave frequencies is promising, the propagation characteristics of these frequencies are very different from those of microwave frequency bands ^[8]. These propagation characteristics include path loss, diffraction and blockage, rain attenuation, air absorption, and foliage loss behaviours. The massive multiple-input multiple-output (MIMO) technology is a potential answer to this problem. With this

technology, a considerable number of antennas might provide sufficient gain to compensate for the route loss ^{[9][10]}. Massive MIMO over mmwaves is a novel approach that combines the benefits of massive MIMO antenna arrays with the vast amount of bandwidth that is currently accessible over mmwaves ^[11]. Massive MIMO over mmwaves has the potential to revolutionise wireless communication. Beam squint is one of the issues that arise in mmwave communications. This issue manifests itself as a change in the spatial direction of a beam as a function of frequency. As a consequence, there are discernible variations in the path phases at various frequencies.

2. Broadband Communication

2.1. Millimeter Waves and Massive MIMO

In order to attain more bandwidth than traditional frequency bands, 5G and the following 6G mobile networks operate in high-frequency zones ^[12]. This allows for high-capacity wireless transmission of data at speeds up to several gigabits per second (Gbps) $\begin{bmatrix} 13 \\ 13 \end{bmatrix}$. It is generally agreed that mmwave communication $\begin{bmatrix} 14 \\ 14 \end{bmatrix}$ is one of the most crucial technologies for reaching peak data rates of 10 Gbit/s [15]. This is due to the fact that increasing the bandwidth is an effective method for increasing the system's capacity [7] and that there is a lot of bandwidth available in the mmwave channels. Measurement and modeling of mmwave channels [16] and assessment of the band's potential for future wireless systems have received significant academic and industrial attention because of the spectrum's abundance [17]. However, due to negative propagation effects, particularly those brought on by obstructions in the line of sight (LoS), these high frequencies have been used only sparingly up to this point ^[18]. It is important to handle the greater propagation losses associated with mmwave frequencies, especially beyond 100 m and in non-line of sight (NLoS) environments ^[19]. On the one hand, signal transmission becomes more difficult in areas with dense foliage and higher propagation loss ^[20]. Nonetheless, the link could be forced into the low signalto-noise ratio (SNR) zone due to the constrained send signal power dispersed over a broad bandwidth ^[21]. To address these problems, several transceiver topologies have been created that steer the received or broadcast beams in the appropriate direction. ^[22]. The two main technologies that will enable 5G mmwave communication are MIMO and beamforming ^[23]. Beamforming is a method for using MIMO/array elements to direct the main lobe radiation beam at the transmitters and receivers in the desired directions, eliminating the undesirable spatial selectivity ^[24]. The beamforming array is widely divided into the following categories: analog, digital, and hybrid beamforming array ^[25]. Figure 1 shows the block diagram of massive MIMO system.



Figure 1. Functional block diagram of massive MIMO system ^[26].

2.1.1. Analog Beamforming

The architecture of analog beamforming is shown in **Figure 2**. A phased array is used to accomplish analog beamforming, and the single RF chain is powered by either an analog-to-digital converter (ADC) in the receiver or a digital-to-analog converter (DAC) in the transmitter. Frequency-up converters, power amplifiers, and other components form a transmitter's RF chain; low-noise amplifiers, frequency-down converters, and other components form a receiver's RF chain ^[27].

The phased array's antenna weights are required to be digitally controllable phase changes. The relative amplitudes of the signal sent into the antennas of the transmitter cannot be adjusted, and the phases of the phase shifters are normally quantized to low resolution ^[28]. A beam is created as a result of the send signal, which is constructive in certain directions and destructive in others. On the basis of precise beam-steering tactics, the phase shifters' phases can be dynamically changed. The receiver has a similar set of features ^[27].



Figure 2. An mmwave MIMO system using analog beamforming ^[29].

2.1.2. Digital Beamforming

The radio frequency (RF) signal in a digital array is transformed into a digital signal at the sub-array or element level, and beamforming is accomplished using a digital signal processor ^[30]. In order to process the control signals for beamforming, a digital signal processor is used, which offers additional flexibility and degrees of freedom for the implementation of effective beamforming algorithms. Due to the method's requirement for a separate RF chain for each antenna element, it consumes a great deal of power and has a complicated architecture ^[31]. **Figure 3** shows the structure of digital beamforming.



Figure 3. Digital beamforming [31].

2.1.3. Hybrid Beamforming

Recently, interest in the hybrid analog/digital beamforming array has increased ^{[32][33]}. This type of array increases the efficiency of analog beamforming while reducing the complexity of digital beamforming ^[34]. The hybrid beamformer is composed of two parts: an analog and a digital component ^[35]. The RF chains that compose the digital portion share a common structure with the many plans under discussion. For each user on each subcarrier, a hybrid beamforming's digital component can be used, just like in traditional fully digital beamforming. In contrast, this is not true of the hybrid beamforming analog network or analog component. Since the transmitted signals for all users are combined by the digital beamformers and analog RF beamforming is a post-IFFT (inverse Fast Fourier Transform) operation, all users and subcarriers in an analog network share the same RF component ^[36]. Hybrid beamforming's fundamental architecture is depicted in **Figure 4**.



Figure 4. Hybrid beamforming ^[29].

Beam squint is a phenomenon caused by the frequency dependence of the magnitude of the phase shift in analog beamforming ^[37]. This can cause issues with wideband communication. The Digital Signal Processing (DSP) logic and digital beamforming algorithms can both use actual time delay. As a result, a phased array architecture with every element digitalized would naturally lend itself to solving the frequency dependency problem as well as offer the greatest degree of programmable flexibility. However, issues with this solution's strength, size, and price can arise.

A time delay among elements occurs as a wavefront approaches an array of elements, and it depends on the wavefront angle in relation to boresight. The beam steering can be achieved with a single frequency by substituting a phase shift for the time delay ^[38]. This is true for narrowband waveforms, but in wideband waveforms when a phase shift is used to provide beam steering, the beam's direction can change with frequency. As a result, the needed phase shift varies with frequency for a given beam direction. Alternately, the beam direction varies with frequency for a given phase shift. Beam squint is the effect where the beam angle changes depending on the frequency ^[39]. **Figure 5** and **Figure 6** shows the effect of beam squint.



Figure 5. The effect of beam squint [9].



Figure 6. Beam squint effect for a phased array antenna using electrical phase shifters operating at frequencies in the range of 10–20 GHz ^[39].

In addition, keep in mind that there is no phase shift between the elements at boresight ($\theta = 0^{\circ}$); hence, there is no way to induce any beam squint. As a result, both the frequency variation and the angle must be functions of the amount of beam squint ^[38]. Beam squint can seriously affect performance since mmwave communications depend heavily on accurate beam alignment between the transmitter and the receiver ^[40].

2.2. Effect of Beam Squint on the Communication System Performance

• Bandwidth and Capacity:

The system bandwidth in wideband communication is constrained by the beam-squint phenomenon ^[41]. The beam squint effect is closely correlated to bandwidth; as bandwidth rises, the beam squint effect becomes more pronounced, and a larger fraction of the subcarriers experience tiny array gains. Even outside the main lobe, certain subcarriers may exist. As a result, the whole system's capacity starts to decline ^[42]. **Figure 7** shows the channel capacity versus the bandwidth with and without beam squint. It is clear from this figure that the capacity decreases as the array's element count, fractional bandwidth, and beam focus angle rise.



Figure 7. Capacity as a function of bandwidth with beam squint (CBS) and without beam squint (CNBS) [41].

Channel estimation:

Channel estimation algorithms for Orthogonal Frequency Division Multiplexing (OFDM), Maximum Mean Square Error (MMSE) estimator, and Maximum Likelihood Estimation (MLE) estimator are widely used ^[41]; however, these estimators do not consider beam squint phenomena. In ^[42], the MLE channel estimator was used to investigate the effect of beam squint on the estimation process, Fading was considered fixed, and the results showed that beam squint increases the error of channel estimation, even without any noise. The error also increases with increasing the number of antenna elements in the array, fractional bandwidth, and the magnitude of beam focus angle. **Figure**

8 shows an example of the MLE channel estimation with beam squint, where H(n) represents the true channel, and $H^{\hat{}}$ (n) represents the estimated channel. It is clear from this figure that there is a difference between the actual channel and the estimated channel, even without considering the noise.



Figure 8. Example of MLE channel estimate using the receiver's beam squint. There is no beam squint in the transmitter and only one path. No more noise exists. The number of antennas $N = 16 \frac{[41]}{10}$.

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