Class D Audio Power Amplifier

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Class D power amplifiers, one of the most critical devices for application in sound systems, face severe challenges due to the increasing requirement of smartphones, digital television, digital sound, and other terminals. The audio power amplifier has developed from a transistor amplifier to a field-effect tube amplifier, and digital amplifiers have made significant progress in circuit technology, components, and ideological understanding. The stumbling blocks for a successful power amplifier are low power efficiency and a high distortion rate.

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1. Introduction

The power amplifier is a type of electronic amplifier designed to increase the magnitude of the power of a given input signal. The audio power amplifier circuit is an essential branch of the power amplifier, one of the essential components of a sound system. It is designed as a final block in an audio amplifier chain to increase the input signal to a level that is high enough to drive loads. It has come a long way under the influence of the rapid development of integrated circuits in recent decades.

The audio power amplifier market is driven by the increasing demand for mobile audio devices, the growing prevalence of consumer electronics devices worldwide, the growing appeal of in-car infotainment systems, and the growing market for high-quality audio devices. According to the report by ^[1], "The audio amplifier market size is expected to grow from USD 3.4 billion in 2019 to USD 4.4 billion by 2024, at a CAGR of 5.6% from 2019 to 2024". A larger market makes upgrading and improving audio amplifiers very meaningful.

The power efficiency comparison of audio amplifiers is shown in **Figure 1**. Traditional analog power amplifiers can be classified into Class A, Class B, Class AB, and other forms according to different working conditions. These kinds of amplifiers have high fidelity, which means they can provide superior sound quality. However, the power conversion efficiency of linear and non-linear amplifiers is extremely low. The emergence of Class D amplifiers solves this problem. The Class D amplifier was first devised in 1958 ^[2]. Compared to linear power amplifiers (Class A) and non-linear amplifiers (Class B, Class B, Class C) ^{[3][4]}, the primary advantage of switching-type Class D amplifiers is their theoretical 100% power efficiency ^[5]. The power efficiency of Class D amplifiers at a maximum modulation index is typically higher than 90% ^[6]. In contrast, the power efficiency of linear amplifiers such as Class B and Class AB power amplifiers was only 70% or less. This is why the Class D amplifier is currently the best choice for commercial audio amplifiers and is gradually replacing linear counterparts ^[7].



Figure 1. Efficiency comparison of each amplifier ^{[2][8][9]}.

The switching device in the output stage is the main point in the history of Class D amplifier development. Germanium transistors were a part of early Class D amplifier design but proved unsuitable for switching topology, hence, early amplifier designs were unsuccessful. Nevertheless, thanks to the advent of MOSFET technology, the relative technology of Class D design was again developed by engineers. The Class D amplifier is currently widely used in various fields such as flat-screen televisions and car sound-control units. Class D amplifiers use a pair of switching devices for push/pull configuration.

The performance of the switching device directly affects the power efficiency and high fidelity of the audio system. Due to the limitation of material characteristics, it is difficult to improve the overall performance of existing siliconbased devices by improving the device structure and manufacturing process ^{[5][10]}. Therefore, new highperformance power devices are needed as substitutes to achieve better conversion efficiency, which requires minor conduction and lower switching losses to suit variable frequency occasions.

2. Topologies for Class D Amplifiers

The amplifier's topologies are the basic structure that depends on the work mode and parameters such as power efficiency and degree of distortion. Researchers analyze the Class D amplifier's topological model to determine the advantages and disadvantages, lists standard amplifier design examples, and provides the basis for improving the amplifier.

A widespread closed-loop analog Class D amplifier covers integrators, modulators, output stage, low-pass filters, and loudspeakers (**Figure 2**). In the first step, integrators and filters provide high loop gain and attenuate unwanted

carrier components in the loop. The low-pass filter is always designed between the output model and the loudspeaker to decrease EMI (electromagnetic interference) and exclude extraordinary frequency energy, which is transformed to [11][12].



Figure 2. Block diagram of an audio amplifier.

The transistors operating in the saturation or linear region limit the power efficiency of a linear amplifier due to the significant voltage drops that are caused by the large current, resulting in a substantial power loss that greatly influences the efficiency. Class D amplifiers have a unique topology that dissipates less power than the linear amplifiers mentioned above. The output stage works by switching between positive and negative and producing a series of voltage pulses, while the output transistor has no current without switching and low V_{DS} when managing the current, thus forming smaller I_{DS} ^[13]. The topologies of Class D amplifiers are shown in **Figure 3**a; analog audio signals are sampled and pulse-width-modulated signals are generated. These signals are used to drive the output switches. The signals generated by the output stage will be filtered by the filtering circuit or the speaker itself and converted into sound signals.



Figure 3. The topology and wave (a) form for Class D amplifier and GaN-based design (b) [14].

The Class D amplifier's performance is outstanding at the medium and high-power output. However, because of the losses in power devices, the efficiency is lowest if the power output is inadequate. Some Class D amplifiers operate in two modes to overcome this challenge. The multistage technique limits the output voltage to which the power device can switch if playing low-volume audio. Once the output volume reaches a set threshold, the output voltage rail of the switch is increased to provide a complete voltage swing. Zero voltage switching (ZVS) technology can be used at a low output and hard switching at high-power levels to help reduce the impact of switching losses ^{[15][16]}.

The output stage is the foundation of the power amplifier. **Figure 4** shows an essential CMOS (complementary metal oxide semiconductor) linear switch output stage, which amplifies the signal current into a large current that can drive loudspeakers by adding the energy of DC to the slight wind to the input through the triode. In traditional transistor amplifiers, the output stage contains a transistor that provides a continuous instantaneous output current. Many audio linear amplifiers contain Class A, Class AB, and Class B amplifiers. Compared to digital amplifiers and Class D amplifiers, the power consumption is significant even in the most efficient linear region. This diversity makes Class D amplifiers hugely beneficial in many applications. Lower power consumption generates less heat, saves space and cost, and extends the service time of mobile systems.



Figure 4. CMOS linear switch output stage.

3. GaN-Based Class D Amplifier

In the case of mobile devices with limited battery capacity, working efficiency is a crucial design metric. Although the Class D amplifier has a power consumption advantage over other amplifiers, it still suffers from low fidelity, mainly due to switching distortion.

The processors of Class D amplifiers produce small high-frequency pulse width modulation (PWM) signals that represent the auditory signal. The power transistor converts small signals into significant signals in half-bridge or

full-bridge configurations and drives the speaker through a filter. Increasing the adoption frequency and the output stage switching frequency can significantly increase the sound because each pulse is a square wave. Power is wasted by switching losses and conduction losses in each switching cycle, thus, balancing sound quality, operating frequency, and power consumption ^[17]. Switching loss is another factor that requires proper consideration. Thus, the essential component of the Class D amplifier is the switching device, whose quality directly determines the strength of each indicator of the amplifier (examples are SNR, THD, and TID (transient intermodulation distortion)). Class D amplifiers must employ integrated feedback circuits to compensate for poor open-loop performance, in order to complete the distortion performance goals (THD + N, TIM, and IM) that are required for high-quality sound.

The source of this distortion is the power MOSFETs based on Si. Thus, the primary goal of a progressive Class D amplifier is to improve its switching structure. An excellent choice to for the MOSFET or IGBT is the GaN device or SiC device [18].

3.1. Advantages of GaN Power Device

After decades of intensive development, Si-based power devices are approaching their material limits in terms of performance. Power devices based on wide-bandgap semiconductors with a higher critical breakdown field are desired to enhance the device performance further, and thus, the power conversion efficiency. Research on power devices has been ongoing for many years. The advent of wide bandgap (WBG) semiconductors based on gallium nitride (GaN) or silicon carbide (SiC) enables Class-D amplifiers to achieve outstanding performance regarding distortion and bandwidth ^[19]. According to ^[20], GaN devices have the advantage of operating at higher voltages and lower leakage currents than Si devices. Compared with the first-generation Si and the second-generation GaAs, the third generation GaN and SiC devices have small on-state resistance, fast switching speed, and high voltage performance. The GaN device has a vacuum saturation rate of 2.8 times that of Si devices, making its onresistance and conduction losses much lower than Si devices. The common junction capacitance of GaN devices enables its switching frequency up to the Mhz level. The bandgap of the GaN device is much smaller than that of the silicon device, which causes its critical breakdown electric field to be as high as 3.3 MV/cm; therefore, it has a higher voltage capacity. The comparison of parameters between the GaN device and Si device is shown in Table 1. The larger bandgap width and the insulating damaged electric field reduce the device's conduction resistance and improve its overall performance. The resulting high-electron saturation rate and high carrier mobility enable the device to work at high speeds ^{[21][22]}. Moreover, GaN transistors have more applications due to showing higher switching dynamics and incorporating fewer parasitics in their packages. The experiments in ^[23] demonstrated that GaN devices can lead to a higher power and lower THD in Class D amplifiers. GaN devices play an essential role in many fields, especially in the improvement of power efficiency. It is feasible that these devices could also initiate a revolution in the field of audio power amplifiers [24][25].

Table 1. Comparison of characteristic parameters for various semiconductor materials.

Characteristic	Unit	Si	SIC	GaN
Energy gap	eV	1.1	3.26	3.49
Electron mobility	$\mathrm{cm}^2/\mathrm{Vs}$	1500	700	2000
Saturated electron velocity	$10^7 {\rm cm/s}$	1	2	2.8
Electric breakdown field	MV/cm	0.4	2	3.3
Thermal conductivity	W/cm*K	1.5	4.5	>1.5
Relative permittivity	εr	11.8	10	9

As shown in **Figure 5**, GaN HEMT transistors are the same as Si MOSFET terminals, with a gate, drain, and source ^[26]. This also indicates that GaN HEMT has similar functions to Si MOSFET and can substitute for power devices in Class D amplifiers. They achieve meager resistance with the help of a two-dimensional electron gas (2DEG) between the gate and the source and can be effectively short-circuited due to the availability of an electronic pool ^[27]. The p-Gan gate will stop conduction while there is no gate bias (VGS = 0 V). Gan HEMT is a bidirectional device, unlike silicon devices. Therefore, a reverse current may occur if the drain voltage is permitted to fall below the source voltage. Gan HEMT transistors have the advantage of a clean switching waveform— mainly, the absence of the bulk diodes commonly found in Si MOSFET, which is responsible for much of the switching noise associated with the PN junction ^[28]. If the amplifier works in zero-voltage switching mode, the switching loss and the power loss can be effectively eliminated because the output transformer is realized by inductance current conversion. At present, most GaN power devices have a planar structure like **Figure 5**. If the substrate in this structure has minor lattice and thermal mismatch with GaN material, the uneven phase denotative growth of GaN can be implemented. GaN power devices comprising Si material for the substrate have mature technology and low cost, which is the optimal solution of Class D amplifier power devices ^{[29][30]}.

Source	Insulating Layer	Gate	Insulating Layer	Drain			
	GaN Cap layer						
AlGaN AIN							
• • •	• •	2DEG	• 1	• • • • •			
GaN							
Buffer							
Si Substrate							

Figure 5. Structure for a Classic GaN HEMT.

3.2. Analysis of GaN Class D Amplifier

Table 2 provides several typical power amplifiers using GaN devices, which are characterized by high power efficiency and low THD. The power density of GaN semiconductors is more than ten times that of the latterly diffused metal oxide semiconductor (LDMOS) transistors used in conventional power amplifiers, with higher power density and efficiency. If researchers use GaN to make IP, they can obtain more power for the same size and a smaller size for the same power. For high power, efficient audio power amplifiers, GaN HEMT using power loss, power density, and response speed are better choices. Compared to traditional Si MOSFETs, GaN HEMTs have a higher rate, cost of arable land, and better performance and thermal efficiency.

Reference	Frequency	Power Efficiency	Modulation Method	THD	Max Ouput (Watt)
[<u>31</u>]	13.56 Mhz	87%	PWM	/	/
[<u>32</u>]	2.3 Mhz	/	PWM	/	/
[<u>33</u>]	100–300 Hz	95%	PWM	0.00%	1.75 w
[<u>34]</u>	6.78 Mhz	80-81%	PWM	/	50
[<u>35]</u>	10 kHz	38 mW loss	SHEPWM	0.48%	/
[<u>36</u>]	40 kHz	97.70%	SPWM Unipolar	0.23%	1001.04
[<u>37</u>]	2 Mhz	85.80%	Sigma-Delta	0.02%	55.27
[<u>38]</u>	12.288 mHz	94–97%	Sigma-Delta	0.50%	100
[<u>39]</u>	4–16 kHz	93%	PWM	0.60%	150
[<u>40]</u>	450 kHz	93%	/	0.13%	40 watt
[<u>41</u>]	25 hz–25 kHz	94%	PWM	0.01%	12.5 watt
[42]	200 KhZ	94%	PWM	0.00%	1.3 kW
[23]	4.8 Mhz	96%	/	1%	10 kw

Table 2. The parameters of GaN Class D Amplifier.

Given the critical indicators of Class D amplifiers, GaN devices have corresponding advantages or disadvantages.

3.2.1. Power Efficiency and Distortion

More than one article has mentioned the advantages of GaN devices over MOSFET and IGBT, especially in ^[40]. Further, in ^[41], the effect of the 60 V GaN device and Si device Class D audio amplifier is compared, and the conclusion is drawn that the power density of the GaN power device is much higher than the Si device. In ^[34] is described the effect of GaN transistors in conventional power amplifiers, achieving over 30% power savings and higher power density and efficiency. Compared with traditional Si MOSFET, GaN FET has higher speed, lower

cost, and thermal efficiency. Although the Class D amplifier designed in ^[37] has an efficiency of only 85.8%, this is because the rated circuit of the transistor used in this design is much larger than what the circuit needs, and higher power can be obtained with a more suitable FET.

The power level of Class D amplifiers aims to achieve accurate large-signal replication from small signal sources while reducing energy loss. Better switching properties of the GaN-based FETs produce waveforms that are closer to the desired ideal waveforms than can be achieved by silicon Mofettes. This is an essential benefit of GaN devices, which can switch up to 1000 times faster than silicon FETs. GaN systems' white paper quantifies this superb audio quality, showing that GaN-based Class D amplifiers can produce total harmonic distortion (THD) as low as 0.004%, compared to 0.015% for silicon products ^{[43][44]}. A PDM 25 W Class D amplifier with GaN and Si devices as output stage is tested and compared in ^[45]. The test results show that when the output power is less than 20 W, the power amplifier's distortion based on the GaN device is 0.15% lower than that of the power amplifier based on the Si device, and the power efficiency is 15% higher. In addition, at lower switching frequencies (360 kHz), the two output stages' distortion performance and power efficiency are similar. At higher switching frequencies (>1.1 Mhz), the GaN output stage has considerable advantages ^[46].

In addition, in zero-voltage switching mode, the output is changed by inductive current commutation. Therefore, any switching loss in the switching device and the resulting power loss can be moved. However, to avoid shoot-through between the two devices, a small blanking delay must be added to ensure that the off-state of the switching period is maintained until the on-state of the next switching period is entered. the output waveform is different from the waveform expected by the PWM output, resulting in audio signal distortion. The blank delay time depends on the output capacitance Coss of the power device used. Compared to Si MOSFETs, GaN transistors have significantly lower Coss, which means that blanking delay time can be minimized, thereby minimizing distortion.

Therefore, power density, power efficiency, and switching frequency advantages indicate that GaN devices will enable a smaller circuit footprint, lower thermal radiation, and long service life, which bridges the direct contradiction between power efficiency and sound quality. However, cost and the lack of a dedicated Class D amplifier integrated circuit are disadvantages of GaN devices.

3.2.2. Thermal Conductivity

Heating is one of the main problems of GaN devices due to the high-power densities. Due to the high conductivity of the SiC, it is a common method of growing GaN on SiC substrate to limit the temperatures of GaN HEMT to no higher than the maximum temperature ^[3]. Some variations reported across the literature, such as the GaN output stage, introduce performance reduction after high temperature, which is an essential disadvantage of GaN devices ^{[46][47]}. However, the growth of GaN involves particular transition layers typically involving alloys of GaN and AIN, which have inferior thermal conductivity because of disorder-induced phonon scattering. In some cases, the performance limiting factor of GaN might be its thermal conductivity as opposed to its electronic properties. Thus, the heat dissipation of audio systems using GaN devices will be one of the following research focuses. The TMT

(Thermal Management Technologies) program introduces several cooling technologies of heat management in semiconductors, such as attached cooling and embedded cooling technologies ^{[48][49][50]}. The junction temperature of the GaN devices is usually generated at the drain side of the gate, and for this reason, the thermal management solutions should focus on the heat area.

In ^{[51][52]}, near-junction embedded microfluidic cooling methods are recommended, which have an excellent effect. This technology is already suitable to replace traditional passive remote cooling techniques.

In ^[52], the feasibility of using diamond as a substrate for GaN devices was investigated. In ^[53], diamond microfluidics-based intrachip cooling was used to influence the heat outflow from a model with low thermal conductivity. In addition, the cooling effects of active and passive cooling techniques on GaN devices with SiC as substrate and diamond as substrate were studied in ^[54].

3.2.3. EMI

The high switching speed of GaN devices will deteriorate the EMI in the amplifier. There have been many articles studying the EMI produced by GaN in converters and its effects, for example [55][56][57][58][59]:

Compared with the Si MOSFET, the full-speed (more than 100 MHz) GaN device increases the EMI to 10 dB. Some EMI reduction schemes are proposed, such as selecting the original ^[60], modifying the topology of the converter circuit ^[59], or improving the filter ^[59]. The research on audio amplifiers is still insufficient at present:

In ^[60], audio amplifiers were built using the SPICE models of fundamental drivers and GaN transistors, which allow an accurate GaN power stage. Moreover, the research hypothesizes that the model and the object may not be accurate; this error occurs in the low-frequency band. The reason is that the simulation model produces significant spikes of drain currents at the beginning of transistors turning on. In ^[61], a simple method for the conductive EMI modeling of integrated Class D amplifiers is shown. However, the work does not consider the control circuits of GaN transistors.

The influence of the transistor's current spikes on the EMI level grows with increasing frequency. Higher switching frequencies flatten the interference spectrum in the entire measurement band for both modulated signals and signals with no modulation. The influence of stray parameters also increases.

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